

4-2 解题过程：

(1) 由  $f(t) = \sin wt [u(t) - u(t - T/2)] = \sin wt u(t) + \sin [w(t - T/2)] u(t - T/2)$  得

$$\begin{aligned} \mathcal{L}[f(t)] &= \mathcal{L}[\sin wt u(t)] + \mathcal{L}\{\sin [w(t - T/2)] u(t - T/2)\} \\ &= \frac{\omega^2}{s^2 + \omega^2} + \frac{\omega}{s^2 + \omega^2} e^{-\frac{sT}{2}} \\ &= \frac{\omega}{s^2 + \omega^2} \left( 1 + e^{-\frac{sT}{2}} \right) \end{aligned}$$

(2) 由  $\sin(wt + \varphi) = \sin wt \cos \varphi + \cos wt \sin \varphi$  得

$$\begin{aligned} \mathcal{L}[\sin(wt + \varphi)] &= \mathcal{L}(\sin wt \cos \varphi) + \mathcal{L}(\cos wt \sin \varphi) \\ &= \frac{\omega \cos \varphi}{s^2 + \omega^2} + \frac{s \sin \varphi}{s^2 + \omega^2} \\ &= \frac{\omega \cos \varphi + s \sin \varphi}{s^2 + \omega^2} \end{aligned}$$

4-3 解题过程：

(1) 由  $f(t) = e^{-(t-2)} u(t-2) \cdot e^{-2}$  得

$$\begin{aligned} \mathcal{L}[f(t)] &= e^{-2} \mathcal{L}[e^{-(t-2)} u(t-2)] \\ &= e^{-2} \cdot \frac{1}{s+1} \cdot e^{-2s} \\ &= \frac{1}{s+1} e^{-2(s+1)} \end{aligned}$$

(2)  $\mathcal{L}[f(t)] = \frac{1}{s+1} e^{-2s}$

(3) 由  $f(t) = e^{-t} u(t) \cdot e^2$  得

$$\mathcal{L}[f(t)] = e^2 \mathcal{L}[e^{-t} u(t)] = \frac{e^2}{s+1}$$

(4) 由  $f(t) = \sin[2(t-1) + 2] u(t-1)$

$$= \cos 2 \sin[2(t-1)] u(t-1) + \sin 2 \cos[2(t-1)] u(t-1) \text{ 得}$$

$$\begin{aligned}\mathcal{L}[f(t)] &= \cos 2 \mathcal{L}\{\sin[2(t-1)]u(t-1)\} + \sin 2 \mathcal{L}\{\sin[2(t-1)]u(t-1)\} \\ &= \frac{2 \cos 2}{s^2 + 4} e^{-s} + \frac{s \sin 2}{s^2 + 4} e^{-s} \\ &= \frac{2 \cos 2 + s \sin 2}{s^2 + 4} e^{-s}\end{aligned}$$

(5) 由  $f(t) = (t-1)u(t-1) - (t-2)u(t-2) - u(t-2)$  得

$$\begin{aligned}\mathcal{L}[f(t)] &= \mathcal{L}[(t-1)u(t-1)] - \mathcal{L}[(t-2)u(t-2)] - \mathcal{L}[u(t-2)] \\ &= \frac{1}{s^2} e^{-s} - \frac{1}{s^2} e^{-2s} - \frac{1}{s} e^{-s} \\ &= \frac{1}{s^2} [1 - (1+s)e^{-s}]\end{aligned}$$

4-4 解题过程:

$$(1) \mathcal{L}^{-1}\left(\frac{1}{s+1}\right) = e^{-t}$$

$$(2) \text{ 由 } \frac{4}{2s+3} = \frac{2}{s+\frac{3}{2}} \text{ 得 } \mathcal{L}^{-1}\left[\frac{4}{2s+3}\right] = \mathcal{L}^{-1}\left[\frac{2}{s+\frac{3}{2}}\right] = 2e^{-\frac{3}{2}t}$$

$$(3) \text{ 由 } \frac{4}{s(2s+3)} = \frac{4}{3}\left(\frac{1}{s} - \frac{1}{s+\frac{3}{2}}\right) \text{ 得 } \mathcal{L}^{-1}\left[\frac{4}{s(2s+3)}\right] = \frac{4}{3} \mathcal{L}^{-1}\left[\frac{1}{s}\right] - \frac{4}{3} \mathcal{L}^{-1}\left[\frac{1}{s+\frac{3}{2}}\right]$$

$$(4) \text{ 由 } \frac{1}{s(s^2+5)} = \frac{1}{5}\left(\frac{1}{s} - \frac{s}{s^2+5}\right) \text{ 得}$$

$$\mathcal{L}^{-1}\left[\frac{1}{s(s^2+5)}\right] = \frac{1}{5} \mathcal{L}^{-1}\left[\frac{1}{s}\right] - \frac{1}{5} \mathcal{L}^{-1}\left[\frac{s}{s^2+5}\right] = \frac{1}{5}(1 - \cos\sqrt{5}t)$$

$$(5) \text{ 由 } \frac{3}{(s+4)(s+2)} = \frac{3}{2}\left(\frac{1}{s+2} - \frac{1}{s+4}\right) \text{ 得}$$

$$\mathcal{L}^{-1}\left[\frac{3}{(s+4)(s+2)}\right] = \frac{3}{2} \mathcal{L}^{-1}\left[\frac{1}{s+2}\right] - \frac{3}{2} \mathcal{L}^{-1}\left[\frac{1}{s+4}\right] = \frac{3}{2}(e^{-2t} - e^{-4t})$$

$$(6) \text{ 由 } \frac{3s}{(s+4)(s+2)} = \frac{6}{s+4} - \frac{3}{s+2} \text{ 得}$$

$$\mathcal{L}^{-1}\left[\frac{3s}{(s+4)(s+2)}\right] = \mathcal{L}^{-1}\left[\frac{6}{s+4}\right] - \mathcal{L}^{-1}\left[\frac{3}{s+2}\right] = 6e^{-4t} - 3e^{-2t}$$

$$(7) \quad \mathcal{L}^{-1}\left[\frac{1}{s^2+1} + 1\right] = \sin t + \delta(t)$$

$$(8) \quad \text{由 } \frac{1}{s^2-3s+2} = \frac{1}{s-2} - \frac{1}{s-1} \text{ 得}$$

$$\mathcal{L}^{-1}\left[\frac{1}{s^2-3s+2}\right] = \mathcal{L}^{-1}\left(\frac{1}{s-2}\right) - \mathcal{L}^{-1}\left(\frac{1}{s-1}\right) = e^{2t} - e^t$$

$$(9) \quad \text{由 } \frac{1}{s(RCs+1)} = \frac{1}{s} - \frac{1}{s+\frac{1}{RC}} \text{ 得 } \mathcal{L}^{-1}\left[\frac{1}{s(RCs+1)}\right] = 1 - e^{-\frac{t}{RC}}$$

$$(10) \quad \text{由 } \frac{1-RCs}{s(RCs+1)} = \frac{1}{s} - \frac{2}{s+\frac{1}{RC}} \text{ 得 } \mathcal{L}^{-1}\left[\frac{1-RCs}{s(RCs+1)}\right] = 1 - 2e^{-\frac{t}{RC}}$$

$$(11) \quad \text{由 } \frac{w}{s^2+w^2} \cdot \frac{1}{(RCs+1)} = \frac{RCw}{1+(RCw)^2} \left( \frac{1}{s+\frac{1}{RC}} - \frac{s}{s^2+w^2} + \frac{\frac{1}{RCw} \cdot w}{s+\frac{1}{RC}} \right) \text{ 得}$$

$$\mathcal{L}^{-1}\left[\frac{w}{s^2+w^2} \cdot \frac{1}{(RCs+1)}\right] = \frac{RCw}{1+(RCw)^2} \left[ e^{-\frac{t}{RC}} - \cos wt + \frac{1}{RCw} \sin wt \right]$$

$$(12) \quad \text{由 } \frac{4s+5}{s^2+5s+6} = \frac{7}{s+3} - \frac{3}{s+2} \text{ 得 } \mathcal{L}^{-1}\left[\frac{4s+5}{s^2+5s+6}\right] = 7e^{-3t} - 3e^{-2t}$$

$$(13) \quad \text{由 } \frac{100(s+50)}{s^2+201s+200} = \frac{100(s+50)}{(s+1)(s+200)} \text{ 得}$$

$$\mathcal{L}^{-1}\left[\frac{100(s+50)}{s^2+201s+200}\right] = \frac{100}{199}(49e^{-t} + 150e^{-200t})$$

$$(14) \quad \text{令 } \frac{s+3}{(s+1)^3(s+2)} = \frac{k_1}{s+2} + \frac{k_2}{(s+1)^3} + \frac{k_3}{(s+1)^2} + \frac{k_4}{s+1}$$

$$\text{则 } k_1 = \left. \frac{s+3}{(s+1)^3} \right|_{s=-2} = -1, \quad k_2 = \left. \frac{s+3}{s+1} \right|_{s=-1} = 2,$$

$$k_3 = \left. \frac{d}{ds} \left( \frac{s+3}{s+1} \right) \right|_{s=-1} = -1, \quad k_4 = \left. \frac{d^2}{ds^2} \left( \frac{s+3}{s+1} \right) \right|_{s=-1} = 1$$

从而 
$$\frac{s+3}{(s+1)^3(s+2)} = \frac{-1}{s+2} + \frac{2}{(s+1)^3} - \frac{1}{(s+1)^2} + \frac{1}{s+1}$$

所以 
$$\mathcal{L}^{-1}\left[\frac{s+3}{(s+1)^3(s+2)}\right] = -e^{-t} + (t^2 - t + 1)e^{-t}$$

(15) 由  $\frac{A}{s^2+K^2} = \frac{A}{K} \cdot \frac{K}{s^2+K^2}$  得  $\mathcal{L}^{-1}\left[\frac{A}{s^2+K^2}\right] = \frac{A}{K} \sin Kt$

(16) 由于  $\mathcal{L}^{-1}\left[\frac{s}{(s^2+3)^2}\right] = \frac{1}{2\sqrt{3}} t \sin \sqrt{3}t$  由拉氏变换的积分性质可得

$$\mathcal{L}^{-1}\left[\frac{1}{(s^2+3)^2}\right] = \int_0^t \frac{1}{2\sqrt{3}} \tau \sin(\sqrt{3}\tau) d\tau = \frac{\sqrt{3}}{18} \sin(\sqrt{3}t) - \frac{t}{6} \cos(\sqrt{3}t)$$

4-19 解题过程：

由于  $f(t)$  可以写作  $f(t) = \sum_{k=0}^{\infty} f_1(t-kT)$

$$= F_1(s) \sum_{k=0}^{\infty} e^{-skT} = \frac{F_1(s)}{1-e^{-sT}}$$

则  $\mathcal{L}[f(t)] = F(s) = \mathcal{L}\left[\sum_{k=0}^{\infty} f_1(t-kT)\right]$

$$= \sum_{k=0}^{\infty} \mathcal{L}[f_1(t-kT)] = \sum_{k=0}^{\infty} F_1(s) e^{-skT}$$

$$= F_1(s) \sum_{k=0}^{\infty} e^{-skT} = \frac{F_1(s)}{1-e^{-sT}}$$

4-20 解题过程：

(1) 周期矩形脉冲信号的第一个周期时间信号为  $f_1(t) = u(t) - u\left(t - \frac{T}{2}\right)$

所以  $F_1(s) = \frac{1}{s} \left(1 - e^{-\frac{T}{2}s}\right)$  则  $F(s) = \frac{F_1(s)}{1-e^{-sT}} = \frac{1 - e^{-\frac{T}{2}s}}{s(1 - e^{-sT})} = \frac{1}{s \left(1 + e^{-\frac{T}{2}s}\right)}$

(2) 正弦全波整流脉冲信号第一周期时间信号为

$$f_1(t) = \sin(wt) \left[ u(t) - u\left(t - \frac{T}{2}\right) \right] = \sin wt u(t) + \sin \left[ w\left(t - \frac{T}{2}\right) \right] u\left(t - \frac{T}{2}\right)$$

$$\text{所以 } F_1(s) = \frac{w}{s^2 + w^2} + \frac{w}{s^2 + w^2} e^{-\frac{T}{2}s} \quad \text{则 } F(s) = \frac{F_1(s)}{1 - e^{-\frac{T}{2}s}} = \frac{w}{s^2 + w^2} \cdot \frac{1 + e^{-\frac{T}{2}s}}{1 - e^{-\frac{T}{2}s}}$$

4-27 解题过程：由  $e(t) = e^{-t}$  得  $E(s) = \mathcal{L}[e(t)] = \frac{1}{s+1}$

$$r_{zs}(t) = r(t) = \frac{1}{2} e^{-t} - e^{-2t} + 2e^{-3t}$$

$$R_{zs}(s) = \mathcal{L}[r_{zs}(t)] = \frac{1}{2(s+1)} - \frac{1}{s+2} + \frac{2}{s-3}$$

故  $H(s) = \frac{R_{zs}(s)}{E(s)}$

$$\begin{aligned} &= \left[ \frac{1}{2(s+1)} - \frac{1}{s+2} + \frac{2}{s-3} \right] \cdot (s+1) \\ &= \frac{1}{2} - \frac{s+1}{s+2} + \frac{2(s+1)}{s-3} \\ &= \frac{3}{2} + \frac{1}{s+2} - \frac{8}{s-3} \end{aligned}$$

所以  $h(s) = \mathcal{L}^{-1}[H(s)] = \frac{3}{2} \delta(t) + (e^{-2t} + 8e^{3t}) u(t)$

4-35 解题过程：

$$H(s) = K \frac{\prod_{i=1}^k (s - z_i)}{\prod_{j=1}^l (s - p_j)} \quad (\text{K 为系数})$$

$$\begin{aligned} &= K \frac{s(s+2-j)(s+2+j)}{(s+3)(s+1-3j)(s+1+3j)} \\ &= K \frac{s(s^2+4s+5)}{(s+3)(s^2+2s+10)} \end{aligned}$$

又知  $H(\infty) = 5$ ，即  $\lim_{s \rightarrow \infty} H(s) = K = 5$

$$H(s) = \frac{5s(s^2+4s+5)}{(s+3)(s^2+2s+10)} = \frac{5(s^3+4s^2+5s)}{s^3+5s^2+16s+30}$$

4-38 解题过程：

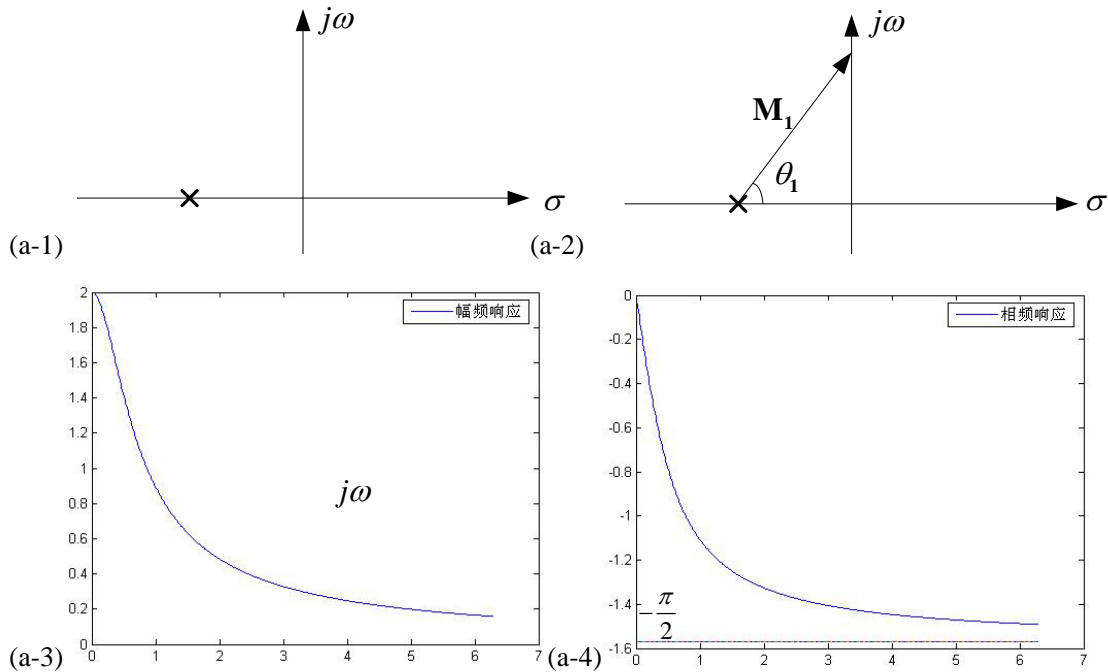
分别画出题图对应的零极点图如下图(a-2)~(f-2)所示。

(1) 由解图(a-1)有，

当  $\omega=0$  时，极点矢量  $\mathbf{M}_1$  最短，辐角  $\theta_1=0$ ，随着  $\omega \uparrow$ ，有  $\mathbf{M}_1 \uparrow$ ， $\theta_1 \uparrow$

当  $\omega \rightarrow \infty$  时， $\mathbf{M}_1 \rightarrow \infty$ ， $\theta_1 \rightarrow \frac{\pi}{2}$

幅频、相频特性如图(a-3)、(a-4)(极点选取-0.5 为例)

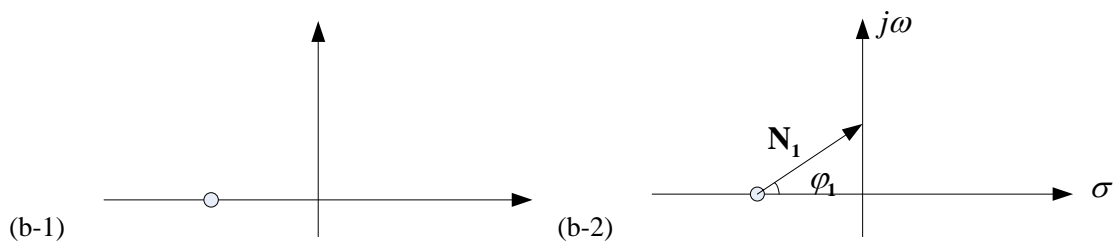


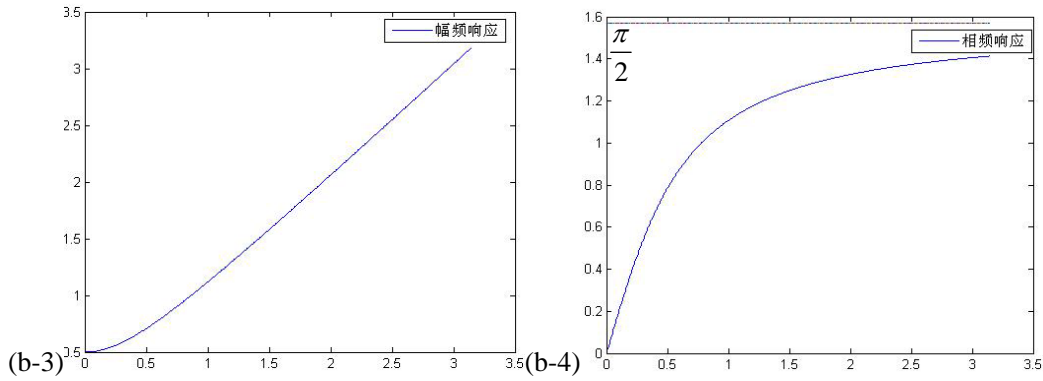
(2) 由解图(b)有，

当  $\omega=0$  时， $\mathbf{N}_1$  最短，辐角  $\varphi_1=0$ ，随着  $\omega \uparrow$ ，有  $\mathbf{N}_1 \uparrow$ ， $\varphi_1 \uparrow$

当  $\omega \rightarrow \infty$  时， $\mathbf{N}_1 \rightarrow \infty$ ， $\varphi_1 \rightarrow \frac{\pi}{2}$

幅频、相频特性如图(以零点为-0.5 为例)





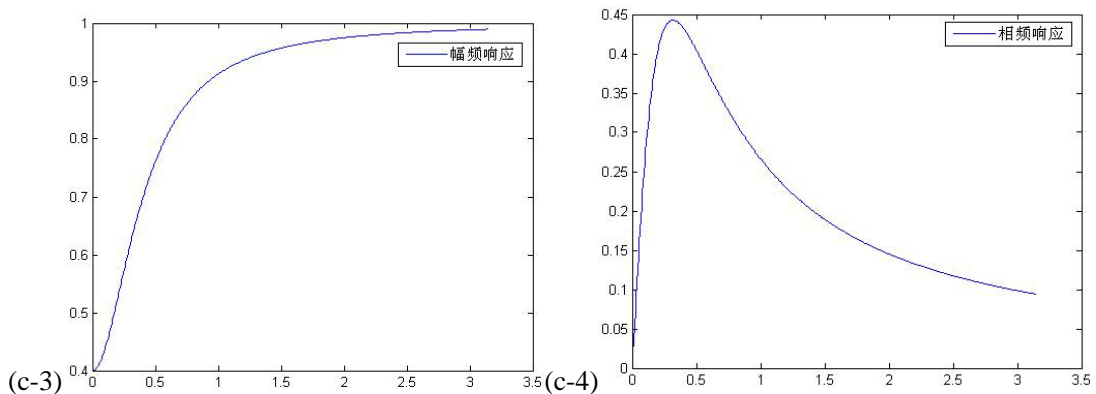
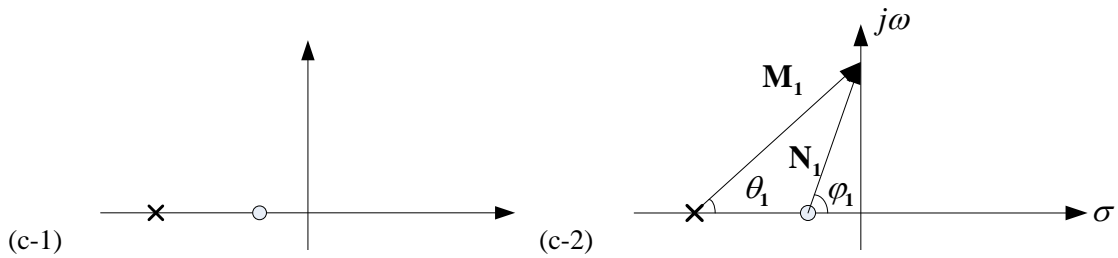
(3) 由解图(b)有,

$\omega = 0$  时,  $\mathbf{M}_1, \mathbf{N}_1$  均为最短, 辐角  $\theta_1 = \varphi_1 = 0$

$\omega \uparrow$ , 则有  $\mathbf{M}_1 \uparrow, \mathbf{N}_1 \uparrow$ , 且有  $\theta_1 \uparrow, \varphi_1 \uparrow$ , 且有  $\theta_1 < \varphi_1$

$\omega \rightarrow \infty$  时,  $\mathbf{M}_1 \rightarrow \infty, \mathbf{N}_1 \rightarrow \infty, \theta_1 \rightarrow \frac{\pi}{2}, \varphi_1 \rightarrow \frac{\pi}{2}$

幅频、相频特性如图(以极点-0.5, 零点-0.2 为例)



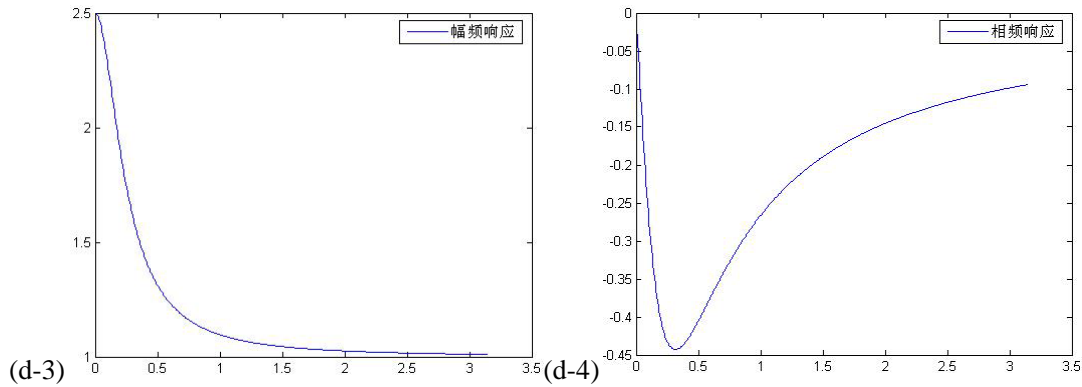
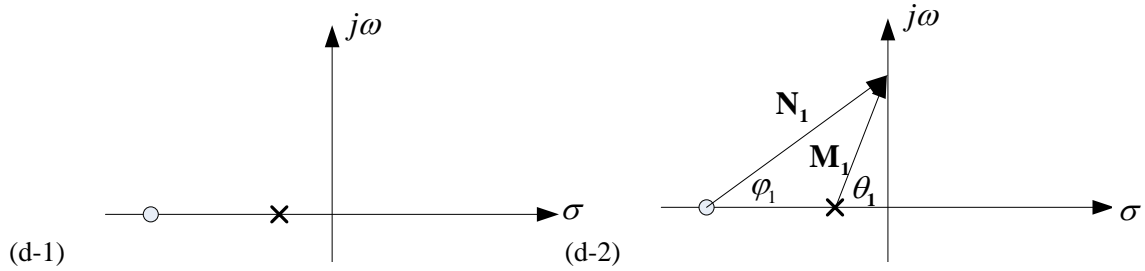
(4) 由解图(d)有,

$\omega = 0$  时,  $\mathbf{M}_1, \mathbf{N}_1$  均为最短, 辐角  $\theta_1 = \varphi_1 = 0$

$\omega \uparrow$ , 则有  $\mathbf{M}_1 \uparrow, \mathbf{N}_1 \uparrow$ , 且有  $\mathbf{M}_1 > \mathbf{N}_1; \theta_1 \uparrow, \varphi_1 \uparrow$ , 且有  $\theta_1 > \varphi_1$

$\omega \rightarrow \infty$  时,  $\mathbf{M}_1 \rightarrow \infty, \mathbf{N}_1 \rightarrow \infty, \theta_1 \rightarrow \frac{\pi}{2}, \varphi_1 \rightarrow \frac{\pi}{2}$

幅频、相频特性如图(以零点-0.5, 极点-0.2 为例)



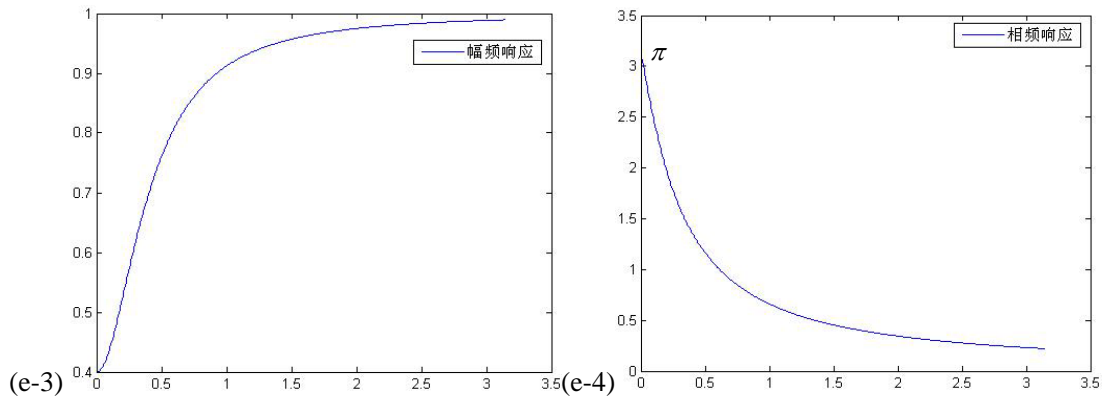
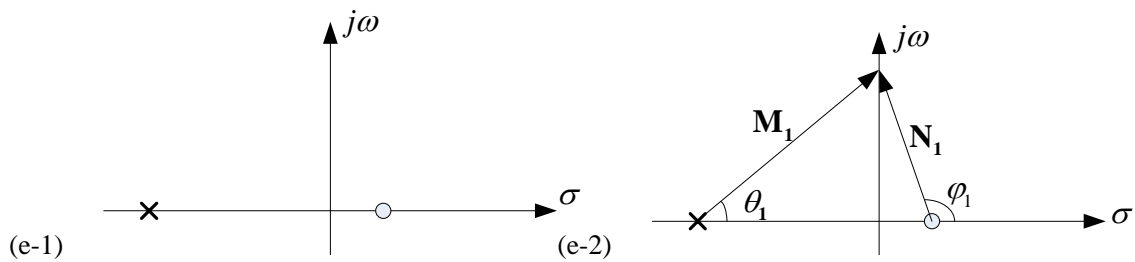
(5) 由解图(e)有,

$\omega = 0$  时,  $M_1, N_1$  均为最短, 辐角  $\theta_1 = 0, \varphi_1 = \pi$

$\omega \uparrow$ , 则有  $M_1 \uparrow, N_1 \uparrow, M_1 > N_1; \theta_1 \uparrow, \varphi_1 \downarrow$

$\omega \rightarrow \infty$  时,  $M_1 \rightarrow \infty, N_1 \rightarrow \infty, \theta_1 \rightarrow \frac{\pi}{2}, \varphi_1 \rightarrow \frac{\pi}{2}$

幅频、相频特性如图(以极点-0.5, 零点 0.2 为例)



(6) 由解图(e)有,

$\omega = 0$  时,  $M_1, N_1$  均为最短, 辐角  $\theta_1 = 0, \varphi_1 = \pi$



$\omega \uparrow$ ，则有  $M_1 \uparrow$ ， $N_1 \uparrow$ ， $M_1 > N_1$ ； $\theta_1 \uparrow$ ， $\varphi_1 \downarrow$ ，但相对关系与(e)中不同。

$\omega \rightarrow \infty$  时， $M_1 \rightarrow \infty$ ， $N_1 \rightarrow \infty$ ， $\theta_1 \rightarrow \frac{\pi}{2}$ ， $\varphi_1 \rightarrow \frac{\pi}{2}$

幅频、相频特性如图(以极点-0.5，零点 0.3 为例)

